Abstract—We consider the problem of tracking the state of Gauss–Markov processes over rate-limited erasure-prone links. We concentrate first on the scenario in which several independent processes are seen by a single observer. The observer maps the processes into finite-rate packets that are sent over the erasure-prone links to a state estimator, and are acknowledged upon packet arrivals. The aim of the state estimator is to track the processes with zero delay and with minimum mean square error (MMSE). We show that, in the limit of many processes, greedy quantization with respect to the squared error distortion is optimal. That is, there is no tension between optimizing the MMSE of the process in the current time instant and that of future times. For the case of packet erasures with delayed acknowledgments, we connect the problem to that of compression with side information that is known at the observer and may be known at the estimator — where the most recent packets serve as side information that may have been erased, and demonstrate that the loss due to a delay by one time unit is rather small. For the scenario where only one process is tracked by the observer–state estimator system, we further show that variable-length coding techniques are within a small gap of the many-process outer bound. We demonstrate the usefulness of the proposed approach for the simple setting of discrete-time scalar linear quadratic Gaussian control with a limited data-rate feedback that is susceptible to packet erasures.

I. INTRODUCTION

Tracking the state of a system from noisy and possibly partially observable measurements is of prime importance in many estimation scenarios, and serves as an important building block in many control scenarios.

The recent rapid growth in wireless connectivity and its ad hoc distributed nature, while offering a plethora of new and exciting possibilities, introduces new design challenges for control over such media. These challenges include, among others, the need to track processes with minimal error over digital links of limited data rate which could be prone to (packet) erasures, and joint processing and reconstruction of distributed processes.

An important scenario, often encountered in practice and depicted in Fig. 1 is that of a multi-track system that tracks several processes over a single shared communication link. In this scenario, at each time instant, several processes are observed by a single observer. The observer, in turn, collects the measured states of these processes into a single vector frame or frame, and maps them into finite-rate packets, which are sent to the state-estimator over a channel which is prone to packet erasures. The state estimator tracks the latest states of the different processes, by constructing minimum mean square error (MMSE) estimates thereof using the available packets received thus far.

Since these settings incorporate communication components, we appeal to relevant tools and results from information theory. The information-theoretic framework for the multi-track setting with a large number of independent processes (large frames) and without packet erasures, was provided by Viswanathan and Berger [1] via the notion of sequential coding for the case of two time steps, and for more steps in [2–5]. In these works the optimal tradeoff between given (per-process) rates and MMSEs (referred to as distortions) were determined when the number of processes is large, in the form of an optimization problem.

A similar control-theoretic framework was also studied by Tatikonda [6–8], who noticed the intimate connection to the works of Gorbunov and Pinsker [9], [10]. For the special case of Gauss–Markov processes, an explicit expression for the achievable sum-rate for given distortions was derived in [2] via the paradigms of predictive coding and differential pulse-code modulation (DPCM) [11] Ch. 6) and extended for the case of three time-steps of independent jointly Gaussian (not necessarily Gauss–Markov) processes, in [12].

In practice, packet-based protocols are prone to erasures and possible delays. The multi-track scenario in the presence of packet erasures was treated under various erasure models. An approach that trades between the performance given all previously sent packets and the performance given only the last packet was proposed in [13]. For random independent identically distributed (i.i.d.) packet erasures, a hybrid between pulse-code modulation (PCM) and DPCM, termed leaky DPCM was proposed in [14] and analyzed for the case of very low erasure probability in [15]. The scenario in which the erasures occur in bursts was considered in [16].

All of these works assume no acknowledgment (ACK) of the arrival status of transmitted packets, i.e., the observer does not know whether transmitted packets successfully arrived to
the state estimator or not.

Nonetheless, in TCP-based networks, packet arrivals are acknowledged via a communication feedback link, in order to robustify the transmission of the overlying data \([17]\).

In this paper, we first consider the multi-track scenario of Gauss–Markov processes, which is defined formally in Sec. [III] We determine the optimal tradeoff between rates and distortions when the number of processes (frame length) is large, in Sec. [IV] Specifically, we show that greedy quantization that optimizes the distortion at each time is also optimal for minimizing the distortion of future time instants. This insight allows us to extend the result to the case where the rate \(r_t\) available for the transmission of the packet at time \(t\) is determined just prior to its transmission.

The packet-erasure channel with instantaneous ACKs can be viewed as a special case of the above noiseless channel with random rate allocation, with \(r_2 = 0\) corresponding to a packet-erasure event \([18]\). The optimal tradeoff between rates and distortions for the multi-track scenario of Gauss–Markov processes in the presence of packet erasures and instantaneous ACKs thereby follows as a consequence.

We further tackle, in Sec. [VI] the delayed ACK setting, in which the observer does not know whether the most recent packets have arrived or not. By viewing these recent packets as side information (SI) that is available at the observer, and possibly at the state estimator, and leveraging the results of Kaspi \([19]\) along with their specialization for the Gaussian case \([20]\), we adapt our scheme of Sec. [III] to the case of delayed ACKs. We provide a detailed description of the proposed scheme for the case where ACKs are delayed by one time unit and demonstrate that the loss due to the delay is small.

In Sec. [VII] we go on and consider the case of tracking a single process, and a variable-length coding scenario (VLC) \([21]\) Ch. 5), in which the packet size is not fixed and is instead constrained to be below a desired rate on average. We consider using a scheme that sequentially applies entropy-coded dithered quantization (ECDQ) \([22]–[24], [25]\) Ch. 5], redolent of the scheme in \([26]\), and we show that it attains an MMSE–rate tradeoff that is close to the large-frame outer bound of Sec. [III]

By supplementing the state tracking task with appropriate control actions in Sec. [VIII] we demonstrate the applicability of the derived results in Secs. [III] and [V] to the scenario of linear quadratic Gaussian (LQG) networked control, where a scalar linear plant driven by an i.i.d. Gaussian process is stabilized by a controller that is not co-located with the observer and is separated from it, instead, by a packet-erasure (and more generally, a random-rate budget) channel. We derive inner and outer bounds, on the optimal linear quadratic regulator (LQR) cost that extend those in \([27], [28]\) to packet-erasure channels. We conclude the paper with Sec. [IX] by discussing the cases of large delays, other types of VLC compression, and tracking of a single process with fixed-length compression.

**II. PROBLEM STATEMENT**

\(|\cdot|\) denotes the Euclidean norm throughout this paper. Random variables are denoted by lower-case letters with temporal subscripts \((\alpha_t, \hat{\alpha}_t)\), and random vectors (frames) of length \(N\) by boldface lower-case letters \((\mathbf{a}, \hat{\mathbf{a}})\). We denote temporal sequences by \(\mathbf{a}^t \triangleq (\alpha_1, \ldots, \alpha_t)\). \(\mathbb{N}\) is the set of natural numbers. All other notations represent deterministic scalars.

The communication spans the time interval \([T] \triangleq \{1, \ldots, T\}, T \in \mathbb{N}\).

We next describe the state dynamics, and the operations carried by the observer and the state estimator operation, which communicate over a finite-rate channel, all of which are also depicted in Fig. [I].

**State dynamics.** Consider \(N\) independent Gauss–Markov processes \(\{s_{t;1}\}, \{s_{t;2}\}, \ldots, \{s_{t;N}\}\) with identical statistics. This can be compactly represented in a vector form as (we assume \(s_0 = 0\) for convenience):

\[
s_t = \alpha_t s_{t-1} + w_t, \quad t \in [T],
\]

where \(s_t \triangleq (s_{t;1}, s_{t;2} \ldots s_{t;N})^T\) is the vector state or frame at time \(t\), \(\{\alpha_t\}\) are known coefficients, the entries of \(w_t\) are the \(N\) independent driving noises the entries of which are i.i.d. Gaussian with zero mean and variance \(W_t\).

Denote the average power of each state at time \(t\) by \(S_t \triangleq \mathbb{E}[s_{t;n}], n \in [N]\). Then, we have the following recursive relation (with \(S_0 = 0\)):

\[
S_t = \alpha_t^2 S_{t-1} + W_t, \quad t = 1, \ldots, T.
\]

**Observer.** Sees the states \(\{s_{t;1}, \ldots, s_{t;N}\}\) of all the \(N\) process at time \(t\), collects them into the frame \(s_t\) and applies a causal function \(\mathcal{O}_t\) to \(s^t\), to generate the packet \(f_t \in \{2^{NR_t}\}\):

\[
f_t = \mathcal{O}_t(s^t),
\]

where \(R_t\) is the per-process rate available for transmission over the channel at time \(t\).

**Channel.** At time \(t\), a packet \(f_t \in \{2^{NR_t}\}\) is sent over a noiseless channel of finite rate \(R_t\),

**State estimator.** Applies a causal function \(\mathcal{D}_t\) to the received packets \(f^t\), to construct an estimate \(\hat{s}_t\) of \(s_t\), at time \(t\):

\[
\hat{s}_t = \mathcal{D}_t(f^t).
\]

**Distortion.** The average quadratic distortion (or MMSE) at time \(t\) is defined as

\[
D_t \triangleq \frac{1}{N} \mathbb{E}[\|s_t - \hat{s}_t\|^2].
\]

In the important special case of fixed parameters,

\[
\alpha_t \equiv \alpha, \quad W_t \equiv W, \quad t \in [T],
\]

the average process power converges to

\[
S_\infty = \frac{W}{1 - \alpha^2},
\]
assuming \(|\alpha| < 1\). In that case, by taking the rate-budget to be fixed too,
\[
R_t \equiv R, \quad t \in [T],
\]
we further define the steady-state distortion (assuming the limit exists):
\[
D_\infty \triangleq \lim_{T \to \infty} D_t.
\]

**Definition (Distortion-rate region).** The distortion–rate region is the closure of all achievable distortion tuples \(D^T \triangleq (D_1, \ldots, D_T)\) for a rate tuple \(R^T \triangleq (R_1, \ldots, R_T)\), for any \(N\), however large; its inverse is the rate–distortion region.

### III. The Distortion–Rate Region

The optimal achievable distortions for given rates for the model of Sec. III are provided in the following theorem.

**Theorem 1 (Distortion–rate region).** The distortion–rate region of Gauss–Markov process multi-track for a rate tuple \(R^T\) is given by all distortion tuples \(D^T\) that satisfy \(D_t \geq D^*_t\) with
\[
D^*_t = (\alpha^2_t D^*_{t-1} + W_t) 2^{-2R_t}, \quad t \in [T],
\]
\[
D^*_0 = 0.
\]

**Remark 1.** The setting of Th. 1 is referred to as “causal encoder–causal decoder” in [2], where for the case of Gauss–Markov processes an explicit expression is provided only for the sum-rate. Torbatian and Yang [12] extend the sum-rate result to the case of three-step general jointly Gaussian processes (not necessarily Markovian). Our work, on the other hand, fully characterizes the rate–distortion region for Gauss–Markov processes.

**Remark 2.** The results and proof (provided in the sequel) of Th. 1 imply that optimal greedy quantization at every step — which is achieved via Gaussian backward [21] Ch. 10.3) or forward [21] pp. 338–339) channels — becomes optimal when \(N\) is large. Moreover, it achieves the optimum for all \(t \in [T]\) simultaneously, meaning that there is no tension between minimizing the current distortion and future distortions.

To prove Th. 1 we first construct the optimal greedy scheme and determine its performance in Sec. III-A. We then show that it is in fact optimal when \(N\) goes to infinity, in Sec. III-B.

#### A. Achievability

We construct an inner bound using the optimal greedy scheme, which amounts to the classical causal DPCM scheme. In this scheme all the quantizers are assumed to be minimum mean square error (MMSE) quantizers, whose quantized values are well known to be uncorrelated with the resulting quantization errors.

**Scheme (DPCM). Observer.** At time \(t\):
- Generates the prediction error
\[
\hat{s}_t \triangleq s_t - \alpha_t \hat{s}_{t-1},
\]
where \(\hat{s}_{t-1}\), defined in (3), is the previous frame reconstruction at the state estimator, and \(\hat{s}_0 = 0\); a linear recursive relation for \(\hat{s}_t\) is provided in the sequel in (11).
- Generates \(\hat{s}_t\), the quantized reconstruction of the prediction error \(\hat{s}_t\), by quantizing \(\hat{s}_t\) using the MMSE quantizer of rate \(R_t\) and frame length \(N\).
- Sends \(f_t = \hat{s}_t\) over the channel.

**State estimator.** At time \(t\):
- Receives \(f_t\).
- Recovers the reconstruction \(\hat{s}_t\) of the prediction error \(\hat{s}_t\).
- Generates an estimate \(\hat{s}_t\) of \(s_t\):
\[
\hat{s}_t = \alpha_t \hat{s}_{t-1} + \hat{s}_t.
\]

**Performance analysis.** First note that the error between \(s_t\) and \(\hat{s}_t\), \(e_t \triangleq s_t - \hat{s}_t\), is equal to \(e_t = s_t - \hat{s}_t\) by (10), (11). Thus, the distortion \(D_t\) is also the distortion in reconstructing \(\hat{s}_t\).

This, along with (1) and (10) means that \(\hat{s}_t = \alpha_t e_{t-1} + w_t\). Since \(w_t\) is independent of \(e_{t-1}\), the average power of the entries of \(\hat{s}_t\) is equal to
\[
\hat{D}_t = \alpha^2_t D_{t-1} + W_t.
\]

Using the property that the rate–distortion function under mean square error distortion of a process with a given average variance is upper bounded by that of an i.i.d. Gaussian process with the same variance (see, e.g., [21] pp. 338–339), we obtain
\[
D_t \leq (\alpha^2_t D_{t-1} + W_t) 2^{-2R_t},
\]
and hence (9) is achievable within an arbitrarily small \(\epsilon > 0\), for a sufficiently large \(N\).

#### B. Impossibility (Converse)

We now prove that, for any frame length \(N \in \mathbb{N}\),
\[
D_t \geq 2^{-2R_t} \mathbb{E}_{f_{t-1}} \left[N \left(s_t | f_t^{-1} = \hat{f}_t^{-1}\right)\right],
\]
\[
D_t \geq D^*_t, \quad t \in [T],
\]
by induction, where the sequence \(\{D^*_t\}\) is defined in (9).
\[
N^*(s_t) \triangleq \frac{1}{2\pi \epsilon} \hat{\mathbb{E}} h(s_t), \quad N^*(s_t | f = \hat{f}) \triangleq \frac{1}{2\pi \epsilon} \hat{\mathbb{E}} h(s_t | f^k - f^t)
\]
denote the entropy-power (EP) and conditional EP of \(s_t\) given \(f^k = \hat{f}^k\), the expectation \(\mathbb{E}_{f_{t-1}} [\cdot]\) is w.r.t. \(f_t^{-1}\), and the random vector \(\hat{f}_t\) is distributed the same as \(f_t\).

**Basic step** \((t = 1)\). Since \(s_0 = 0\), and the vector \(w_1\) consists of i.i.d. Gaussian entries of variance \(W_1\), (12b) is satisfied with equality. To prove (12a), we use the fact that the optimal achievable distortion \(D_1\) for a Gaussian process \((s_1 = w_1)\) with i.i.d. entries of power \(W_1\) and rate \(R_1\) is dictated by its rate–distortion function [21] Ch. 10.3.2]:
\[
D_1 \geq W_1 2^{-2R_1}.
\]

**Inductive step.** Let \(k \geq 2\) and suppose (12) is true for \(t = k - 1\). We shall now prove that it holds also for \(t = k\).

\[
D_k = \mathbb{E} \left[\|s_k - \hat{s}_k\|^2 | f^{k-1}\right]
\]
\[
\geq \mathbb{E} \left[\|s_k - \hat{s}_k\|^2 | f^{k-1} = \hat{f}^{k-1}\right]
\]
\[
\geq \mathbb{E} f_{k-1} \left[\mathbb{E} \left[\|s_k - \hat{s}_k\|^2 | f^{k-1} = \hat{f}^{k-1}\right) 2^{-2R_k}\right]
\]
where (13a) follows from (5) and the law of total expectation, (13h) is by the induction hypothesis, and (13i) holds by the

\[ \mathbb{E}_f|f_{k-1}|^{f_k-1} = f_{k-1} \]

for the Gaussian case with \( D \), with \( \mathbb{E} \) being the expectation operator.

**Assertion 1** (Outer bound for non-Gaussian noise). Consider the setting of Sec. [7] with independent non-Gaussian noise entries \( \{w_t|t \in [T], n \in [N]\} \). Then, the average achievable distortion \( D_t \) at time \( t \in [T] \) is bounded from below by \( D_t \geq D_t^* \), with \( D_0^* = 0 \) and \( D_t^* \) given by the recursion

\[ D_t^* = \alpha^2 D_{t-1}^* + \mathcal{N}(w_t) \]

Proof: The proof is identical to that of the lower bound for the Gaussian case with \( W_t \) replaced by \( \mathcal{N}(w_t) \).

**C. Fixed-Parameter Gauss–Markov Processes**

For the case of fixed parameters \( \{s_t\} \) and fixed rate, the steady-state average distortion is as follows.

**Corollary 1** (Steady state performance with fixed-rate budget). Assume a fixed-parameter \( \{s_t\} \) fixed-rate budget \( \{R_t\} \) setting. If \( \alpha^2 < 2^{-2R} \) then the steady-state distortion is given by

\[ D_{\infty}^* = \lim_{t \to \infty} D_t^* = \frac{W_t^2}{1 - \alpha^2} \]

and is otherwise unbounded.

Proof: The proof is immediate by noting that \( \mathcal{N}(w_t) \) constitutes a linear time-invariant (LTI) system and therefore is globally exponentially stable if the (only) pole of its transfer function lies strictly inside the unit circle, i.e., \( \alpha^2 < 2^{-2R} \), and is unstable otherwise \([29]\) Ch. 6).

**Remark 3.** The result of Cor. [1] remains true for any \( D_0^* \).

Interestingly, the optimal steady-state distortion achievable with a fixed-rate budget \( \{R_t\} \) is in fact optimal even if we loosen this restriction to a total rate-budget constraint. This is a simple corollary of Th. [3] and is formally proved in \([30]\). The same conclusion holds if the frame entries are correlated Gaussians, as was recently proved by Tanaka \([31]\).

**Corollary 2** (Steady state performance with total-rate budget). Let the rates \( R_t \) be subject to a total rate-budget constraint:

\[ \frac{1}{T} \sum_{t=1}^{T} R_t \leq R \]

and define the average-stage distortion by \( D_T = \frac{1}{T} \sum_{t=1}^{T} D_t \). Then, the average-stage steady-state distortion

\[ D_{\infty} = \limsup_{T \to \infty} D_T \]

under a total rate-budget constraint \( \{R_t\} \) is bounded from below by \( D_{\infty} \leq D_{\infty}^* \). Consequently, the fixed (a.k.a. uniform) rate allocation \( R_t = \frac{R}{T} \) is optimal in the limit of \( T \to \infty \).

**IV. RANDOM-RATE BUDGETS**

In practice, the available transmission rate may vary across time depending on the quality of service offered by the infrastructure, as well as, due to other applications sharing the same infrastructure. We therefore generalize next the results of Sec. III to random rates \( \{r_t\} \) that are independent of each other and of \( \{w_t\} \). The rate \( r_t \) is revealed to the observer just before the transmission of time \( t \).

**Theorem 2** (Distortion–rate region). The distortion–rate region of Gauss–Markov multi-track with independent rates \( r_t \) is given by all distortion tuples \( D_T \) that satisfy \( D_t \geq D_t^* \) with

\[ D_t^* = \alpha^2 D_{t-1}^* + W_t \mathbb{E} [2^{-2r_t}] \]

and define the average-stage distortion by \( D_T = \frac{1}{T} \sum_{t=1}^{T} D_t \). Then, the average-stage steady-state distortion

\[ D_{\infty} = \limsup_{T \to \infty} D_T \]

under a total rate-budget constraint \( \{R_t\} \) is bounded from below by \( D_{\infty} \leq D_{\infty}^* \). Consequently, the fixed (a.k.a. uniform) rate allocation \( R_t = \frac{R}{T} \) is optimal in the limit of \( T \to \infty \).

**Remark 4.** An immediate consequence of this theorem and Jensen’s inequality is that using packets of a fixed rate equal to \( \mathbb{E} [r_t] \) performs better than using random rates.

The proof follows by noting that the achievability of Th. [3] is greedy and therefore does not require knowledge of future rates, whereas the impossibility follows by revealing all the rates \( r_t \) to the observer and the state estimator, as a genie, before transmission begins. See \([30]\) for a formal proof.

For the special case of fixed-parameters \( \{s_t\} \) and i.i.d. rates \( \{r_t\} \), the steady-state distortion is given as follows.

**Corollary 3** (Steady state). Assume a fixed-parameter \( \{s_t\} \) setting with i.i.d. rates \( \{r_t\} \). If \( \alpha^2 < 1 \) \([6]\) where \( B \equiv \mathbb{E} [2^{-2r_t}] \), then the steady-state distortion is given by

\[ D_{\infty} = \limsup_{t \to \infty} D_t \]

where \( B = \frac{BW}{1 - \alpha^2} \).

\[^a\text{Recall that in the Gaussian setting } \mathcal{N}(w_t) = \frac{\text{Var}(w_t)}{N} \equiv W_t.\]

\[^b\text{This is trivial for } |\alpha| < 1.\]

\[^c\text{Again, this is trivial for } |\alpha| < 1.\]
Proof: The proof is identical to that of Cor. 1 with $2^{-R}$ replaced by $B$.

V. PACKET ERASURES WITH INSTANTANEOUS ACKS

A. One Packet Per Frame

An important scenario encompassed by the model of Sec. IV is that of packet erasures [18]. Since a packet erasure at time $t$ can be viewed as $r_t = 0$, and assuming that the observer sends packets of fixed rate $R$ and is cognizant of any packet erasures instantaneously, the packet erasure channel can be cast as the random rate channel of Sec. IV with

$$r_t = R b_t = \begin{cases} R, & b_t = 1 \\ 0, & b_t = 0 \end{cases}$$

(20)

where $\{b_t\}$ are the packet-erasure events, such that $b_t = 1$ corresponds to a successful arrival of the packet $f_t$ at time $t$, and $b_t = 0$ means it was erased. We further denote by $g_t \triangleq b_t f_t$ the received output where $g_t = 0$ corresponds to an erasure, and otherwise $g_t = f_t$. We assume that $\{b_t\}$ are i.i.d. according to a $Ber(\beta)$ distribution for $\beta \in [0, 1]$.

**Remark 5.** We shall concentrate on the case of packets of fixed rate $R$ to simplify the subsequent discussion. This way the only randomness in rate comes from the packet-erasure effect. Nevertheless, all the results that follow can be easily extended to random/varying rate allocations to which the effect of packet erasures $\{b_t\}$ is added in the same manner as in (20).

**Corollary 4** (Distortion–rate region). The distortion–rate region with i.i.d. $Ber(\beta)$ packet erasures and instantaneous ACKs is given as in Th. 2 with

$$B \triangleq \mathbf{E}[2^{-2r_t}] = 1 - \beta (1 - 2^{-2R}).$$

(21)

**Corollary 5** (Steady state). The steady-state distortion is given as in Cor. 3 with $B$ as in (21).

**Remark 6.** In contrast to the scenario without packet erasures, the uniform rate allocation can be improved by allowing a dynamic rate allocation that depends on the pattern of packet erasures $b^t$. See Sec. IX-F for a further discussion.

B. Multiple Packets Per Frame

In Sec. V-A we assumed one packet ($f_t$) was sent per each frame ($s_t$). Instead, one may choose to transmit multiple packets of lower rate per one frame. If we assume that each packet arrival is instantly acknowledged, then the resulting scenario falls again in the random-rate framework of Sec. IV.

Specifically, assume that the observer uses $K$ packets of equal rate $R/K$ (and hence a total rate of $R$) to successively refine [32, Ch. 13.5] a single state frame $s_t$. Then, the rate probability distribution amounts to $r_t = R b_t / K$, with $b_t$ denoting the number of successful packet arrivals at time $t$, corresponding to $s_t$. Assuming that the erasure events of all packets are i.i.d. with probability $1 - \beta$ implies that $\{b_t\}$ are i.i.d. according to a Binomial distribution $Bin(K, \beta)$.

Interestingly, the optimal number of packets $K$ depends on the (total) rate $R$ and packet-arrival probability $\beta$, since by allocating more lower-rate packets, one trades a lower probability of receiving the maximal available rate at the state estimator with a higher probability of receiving intermediate rates. The optimal $K$ is determined by the number that minimizes $\mathbf{E}[2^{-r_t}]$, as is demonstrated in Fig. 2.

We note that in absence of ACKs of intermediate packets, the successive refinement encoding considered here cannot be used. One could use repetition coding to trade multiplexing gain with diversity [33] or multiple description coding [34], when ACKs are sent only after all the intermediate packets are transmitted. We do not discuss such extensions in this paper due to a lack of space.

**Remark 7.** We only considered uniform rate allocations for all the packets. Clearly, one can generalize the same approach to non-uniform packet rates.

**Remark 8.** In practice one might expect longer packets to be prone to higher erasure probability. This can be taken into account when deciding on the $K$ that minimizes $\mathbf{E}[2^{-r_t}]$.

VI. PACKET ERASURES WITH DELAYED ACKS

We now tackle the case of i.i.d. packet erasures with ACKs that are delayed by one time unit, i.e., the case where at time $t$ the observer does not know whether the last packet arrived or not (i.e., it does not know $b_{t-1}$), but knows the erasure pattern of all preceding packets (knows $b^{t-2}$). The observer (3) and state estimator (4) mappings can be written as [recall the definition of $g_t \triangleq b_t f_t$]

$$f_t = \mathcal{B}_t \left( s^t, g^{t-2} \right), \quad \hat{s}_t = \mathcal{D}_t \left( g^t \right).$$

To construct a transmission scheme for this case, we recall the following result by Perron et al. [20, Th. 2], which is a specialization to the jointly Gaussian case of the result by Kaspi [19, Th. 1], who established the rate–distortion region of lossy compression with two-sided SI where the SI may or may not be available at the state estimator.

**Theorem 3** ([20, Th. 2]). Let $s$ be an i.i.d. zero-mean Gaussian process of power $S$, which is jointly Gaussian with SI $y$, which

\[^3\]We use a backward channel to represent the SI $s = y + z$, as opposed to the forward channel $y = s + z$ used in [20, 55.]

\[^4\]Kaspi’s result [19, Th. 1] can also be viewed as a special case of [35] with some adjustments; see [36].
is available at the observer and satisfies \( s = y + z \), where \( z \) is an i.i.d. Gaussian noise of power \( Z \) that is independent of \( y \). Denote by \( \hat{s}^+ \) and \( \hat{s}^- \) the reconstructions of \( s \) with and without the SI \( y \), and by \( D^+ \) and \( D^- \) — their mean squared error distortion requirements, respectively. Then, the smallest \( y \) of \( z \) is an i.i.d. Gaussian noise of power \( \log \frac{\Delta}{a+b} \), where \( a\|b \triangleq \frac{\alpha b}{a+b} \) denotes the harmonic mean of \( a \) and \( b \), and \( \Delta \triangleq \sqrt{(S-Z)(S-D^-)}|D^+ - \sqrt{(Z-D^-)(D^- - D^-)}|S| \sqrt{Z} (S-D^-) \).

In our case, at time \( t \), the previous packet \( f_{t-1} \) serves as the SI. Note that this SI is always available to the observer; the state estimator may or may not have access to it, depending whether the previous packet arrived or not. Since the ACK is delayed, during the transmission of the current packet \( f_t \), the observer does not know whether the previous packet was lost.

The tradeoff between \( D^+ \) and \( D^- \) for a given rate \( R \) will be determined by the probability of a successful packet arrival \( \beta \).

**Scheme (Kaspi-based).**

**Observer.** At time \( t \):
- Generates the prediction error \( \hat{s}_t \triangleq s_t - \alpha_t f_{t-1} \hat{s}_{t-2} \).
- Generates \( f_t \) by quantizing the prediction error \( \hat{s}_t \) as in Th. 3 where \( f_t \) is available as SI at the observer and possibly at the state estimator (depending on \( b_{t-1} \)) using the optimal quantizer of rate \( R \) and frame length \( N \) that minimizes the distortion averaged over \( b_{t-1} \):
  \[
  D^\text{Weighted}_t = \beta D^+_t + (1 - \beta) D^-_t ; \quad (22)
  \]
  more precisely, since the observer does not know \((b_{t-1}, b_t)\) at time \( t \):
  - Denote the reconstruction of \( \hat{s}_t \) at the state estimator from \( f_t \) and \( g^{t-1} \) — namely given that \( b_t = 1 \) — by \( Q_t(\hat{s}_t) \), and the corresponding distortion, averaged over \( b_{t-1} \), by \( D^\text{Weighted}_t \).
  - Denote the reconstruction from \( f_t \) and \( g^{t-2} \) — namely given that \( b_0 = 1 \) and \( b_{t-1} = 0 \) — by \( Q_t(\hat{s}_t) \), and the corresponding distortion by \( D^-_t \).
  - Denote the reconstruction from \( f_{t-1} \) and \( f_t \) and \( g^{t-2} \) — namely given that \( b_t = 1 \) and \( b_{t-1} = 1 \) — by \( Q_t(\hat{s}_t) \), and the corresponding distortion by \( D^+_t \).

Then, the observer sees \( \alpha_t Q_{t-1}(\hat{s}_{t-1}) \) as possible SI available at the state estimator to minimize \( D^\text{Weighted}_t \) as in (22).
- Sends \( f_t \) over the channel.

**State estimator.** At time \( t \):
- Receives \( g_t \).
- Generates a reconstruction \( \hat{s}_t \) of the prediction error \( \hat{s}_t \):
  \[
  \hat{s}_t = \begin{cases} 
  Q^+_t(\hat{s}_t), & b_t = 1, b_{t-1} = 1 \\
  Q^-_t(\hat{s}_t), & b_t = 1, b_{t-1} = 0 \\
  0, & b_t = 0
  \end{cases} \quad (23)
  \]
  - Generates an estimate \( \hat{s}_t \) of \( s_t \); \( \hat{s}_t = \alpha_t \hat{s}_{t-1} + \hat{s}_t \).

This scheme is the optimal greedy scheme whose performance is stated next, in the limit of large \( N \).

**Theorem 4.** The following distortions \( D^T \) can be approached arbitrarily closely in the limit \( N \to \infty \) for \( t \in [2, T] \):

\[
D_t = \begin{cases} 
D^+_t, & b_t = 1, b_{t-1} = 1 \\
D^-_t, & b_t = 1, b_{t-1} = 0 \\
\alpha_t D_{t-1} + W, & b_t = 0
\end{cases}
\]

where \( D^+_t \) and \( D^-_t \) are the distortions that minimize (22), 7 such that the rate of Th. 3 satisfies

\[
R^K_\text{Kaspi}(\alpha_t D_{t-1}^+ + W, \alpha_t D_{t-1}^+ + W, D^-_t, \Delta) = R.
\]

The proof is again the same as that of Ths. 1 and 2 with \( \hat{s}_t \) generated as in (23).

**Remark 9.** Here, in contrast to the case of instantaneous ACKs, evaluating the distortions \( \{D_t\} \) in explicit form (recall Cor. 2) is more challenging. We do it numerically, instead.

Somewhat surprisingly, the loss in performance of the Kaspi-based scheme due to the ACK delay is rather small compared to the scenario in Sec. V where the ACKs are available instantaneously, for all values of \( \beta \). This is demonstrated in Fig. 3 where the performances of these schemes are compared along with the performances of the following three simple schemes for \( \alpha_t \equiv 0.7, W \equiv 1, \beta = 0.5, R = 2 \):
- **No prediction:** A scheme that uses no prediction at all, as if the state frames were independent. This scheme

\[
D_{\text{for instant FB}} \quad \text{and} \quad D_{\text{instant FB}}
\]

7For \( \beta \) values close to 0 or 1, the loss becomes even smaller as in these cases using the scheme of Sec. V that assumes that the previous packet arrived or was erased, respectively, becomes optimal.
achieves a distortion of \( D_t = \beta S_t 2^{-2R} + (1 - \beta) S_t \), where \( S_t \) is the power of the \( t \)-th element of \( s \), as given in (2).

- **Assumes worst case (WC):** Since at time \( t \) the observer does not know \( b_t \), a “safe” way would be to work as if \( b_t = 0 \). This achieves a distortion of 
  \[
  D_t = [\alpha^2 D_{t-1} + (1 + \alpha^2)W] [\beta 2^{-2R} + (1 - \beta)] + \beta (1 - \beta) [\alpha^2 D_{t-1} + W], \quad t = 2, \ldots, T, \]
  \[ D_0 = 0, \quad D_1 = W 2^{-2R}. \]

- **Assumes best case (BC):** The optimistic counterpart of the previous scheme is that which always works as if \( b_t = 1 \). This scheme achieves a distortion of 
  \[
  D_t = \beta [\alpha^2 D_{t-1} + W] [2^{-2R} + (1 - \beta)] + (1 - \beta) [\alpha^2 D_{t-1} + W], \quad t = 2, \ldots, T, \]
  \[ D_{t-1} = \alpha^2 D_{t-2} + W, \quad t = 2, \ldots, T, \]
  \[ D_0 = 0, \quad D_1 = W 2^{-2R}. \]

### VII. VARIABLE-LENGTH CODING

In contrast to previous sections where at time instant \( t \) exactly \( N R_t \) bits were available for the compression of the \( N \)-length vector \( s_t \), in this section, we consider the less restrictive case, commonly referred to as VLC, where the (transmit) rate is constrained to \( R \) only on average across time. We assume again a packet-erasure case, where, as in Sec. [V-A] the packet at time \( t \) is erased with probability \( 1 - \beta \), and successfully arrives with probability \( \beta \). The packet-erasure events \( \{b_t\} \) take values in \( \{0, 1\} \) where 0 corresponds to an erasure and 1 — to a successful arrival; we assume that these events are i.i.d.

We further concentrate on the scalar case, \( N = 1 \). The rate constraint can be therefore written as:

\[
\mathbb{E}[r_t|b_t = 1] \leq R, \quad \mathbb{E}[r_t|b_t = 0] = 0, \quad t \in [T],
\]
where, in contrast to previous sections, in this section, \( r_t \) can depend on the exact value of \( s_t \).

**Remark 10.** Similarly to the treatment in Sec. [V-B] the treatment in this section can be extended to the case of multiple packets per state frame.

We first note that the lower bound of Th. [2] remains valid for the VLC case, since Shannon’s classical rate–distortion theorem [37]–[39] extends to the case of VLC (see, e.g., [40]). We next prove that this lower bound can be closely met by incorporating ECDQ [22]–[24], [25] Ch. 5, which is described as follows.

**Scheme (ECDQ).**

- **Offline.** The observer and the state estimator generate a common random dither \( z \) that is uniformly distributed over \( [-\Delta/2, \Delta/2] \).

- **Observer.** Uses a uniform-grid (one-dimensional lattice) quantizer with quantization step \( \Delta \) to quantize \( \gamma s + z : Q_\Delta(\gamma s + z) \), where \( \gamma \) is a pre-determined scalar.

- **Applies entropy coding to the output of the quantizer.**

- **State estimator.** Receives the coded bits.

- **Reconstructs the output of the quantizer:** \( Q_\Delta(\gamma s + z) \).

- **Generates the state estimate by subtracting \( z \) from the quantizer’s output and multiplies the result by \( \gamma \):**

\[
\hat{s} = \gamma [Q_\Delta(\gamma s + z) - z].
\]

**Theorem 5 (ECDQ performance [24], [25] Ch. 5).** The average rate \( R \) needed by the ECDQ scheme (for \( N = 1 \)) to achieve a distortion \( D_t \) for a state \( s \) with variance \( \gamma_s \) and \( \gamma \) set to \( \gamma = \sqrt{1 - D/S} \) is bounded from above by

\[
R \leq \frac{1}{2} \log \frac{\gamma_s}{D} + 1 + \frac{2\pi e}{12},
\]

where the first element in (26) is the Gaussian rate–distortion function and the second element is the “shaping loss”.

Equivalently, the average distortion \( D \) of ECDQ under an average rate constraint \( R \) is bounded from above by

\[
D \leq \frac{2\pi e}{12} \gamma_s^2 2^{-2R}.
\]

**Remark 11 (One-to-one source coding).** The entropy coding employed here is assumed to be one-to-one, that is, we do not require the resulting code to be prefix free. For a more thorough discussion of one-to-one versus prefix-free coding and the rationale behind using each, see Sec. [IX-C].

**Remark 12 (ECDQ for \( N > 1 \)).** For \( N > 1 \), one may replace the uniform scalar quantizer with a lattice-based one; the resulting distortion in this case is bounded from above by

\[
R \leq \frac{1}{2} \log \frac{\gamma_s}{D} + 1 + \frac{2\pi e G_N}{12},
\]

where \( G_N \) is the normalized second moment of the lattice. For the special case of a scalar lattice, \( G_1 = 1/12 \). It is known, by the isoperimetric inequality, that \( G_N \geq \frac{1}{2\pi e} \) for any lattice of any dimensions \( N \). Furthermore, it is known that a sequence of lattices of growing dimensions can be devised that attains this isoperimetric lower-bound in the limit \( N \to \infty \). See [25] for a thorough account of lattices and their application to ECDQ.

We next incorporate ECDQ in the DPCM scheme of Sec. [III-A] we apply ECDQ (with i.i.d. dither \( z_t \) across time) to \( \hat{s}_t \) at the observer and recover it at the state estimator; the rest of the scheme remains exactly the same. The performance of this scheme is stated in the following theorem.

**Theorem 6 (ECDQ-based DPCM scheme performance).** The ECDQ-based DPCM scheme (for \( N = 1 \)) under an average rate constraint \( R \) achieves a distortion \( D_t \) at time \( t \) that satisfies the recursion:

\[
D_t \leq \frac{2\pi e}{12} B (\alpha_2^2 D_{t-1} + W_t),
\]

with \( D_0 = 0 \) and \( B \) as in (21).

Th. 6 suggests that the gap in performance of scalar systems compared to their \( N \)-dimensional counterparts is bounded by a multiplicative factor of \( 2\pi e/12 \) in each recursive step (28).

**Remark 13 (ECDQ-based DPCM scheme for \( N > 1 \)).** Following Rem. 12 for the case of \( N > 1 \) the resulting distortion when applying ECDQ for \( N > 1 \) with an \( N \)-dimensional lattice is bounded from above by

\[
D_t \leq \frac{2\pi e G_N}{12} (\alpha_2^2 D_{t-1} + W_t),
\]
where again $D_0 = 0$, $G_N$ is the normalized second moment of the lattice and $B$ is given in (21).

In the limit of large $T$ we attain the following steady-state distortion.

**Corollary 6** (ECDQ-based DPCM scheme in steady-state). The steady-state distortion of the ECDQ-based DPCM scheme (for $N = 1$) under an average rate constraint $R$ (25) is bounded from above by

$$D_{\infty} \leq \frac{2\pi^2}{12} WB \frac{1 - \frac{2\pi^2}{12} N^2 B}{2}$$  \hspace{1cm} \text{(29)}$$

where $B$ is given in (21).

**VIII. APPLICATION TO NETWORKED CONTROL**

An important application of state tracking is to networked control, namely, to the scenario where, in contrast to traditional control, the observer is not co-located with the controller, and communicates with it instead via a noiseless (“packeted”) channel. Hence, the controller assumes the additional role of the state estimator.

We concentrate on the following simple setting, also depicted in Fig. 4. The channel is the noiseless random-rate budget channel of Sec. IV.

We consider a stochastic system with linear scalar plant evolution which is the same as in (11) (with $s_0 = 0$):

$$s_t = \alpha s_{t-1} + u_t + u_{t-1},$$

where the coefficient $\alpha$ (which is usually assumed to be fixed across time in control applications) can be greater than 1 in its absolute value, corresponding to an unstable open-loop process, with the additional term $u_{t-1}$ serving as the control action that is generated by the controller from all past packets $f_t$, and is used to stabilize the system.

We consider the random-rate budget scenario of Sec. IV.

The goal of the system is to minimize the average-stage LQR cost upon reaching the horizon $T$:

$$\bar{J}_T \triangleq \frac{1}{T} \mathbb{E} \left[ \sum_{t=1}^{T-1} \left( Q_t s_t^2 + R_t u_t^2 \right) + Q_s s_T^2 \right],$$  \hspace{1cm} \text{(30)}$$

where $\{Q_t\}$ and $\{R_t\}$ are known non-negative scalars, respectively, that penalize the cost for state deviations and control actuations, respectively.

In order to derive bounds on the LQR cost for this setting, we use a result by Fischer [41] and by Tatikonda et al. [8], that extends the celebrated control-theoretic separation principle to networked control systems.

**Lemma 1** ([8], [41]). The optimal controller is given by

$$u_t = -K_t \hat{s}_t,$$

where $\hat{s}_t \triangleq \mathbb{E} [s_t | f_t^t]$, $K_t$ is the optimal LQR control gain

$$K_t = \frac{L_{t+1}}{R_t + L_{t+1}},$$

and $L_t$ satisfies the dynamic backward Riccati recursion [42]:

$$L_t = Q_t + \alpha R_t K_t,$$

with $L_{T+1} = 0$. Moreover, this controller achieves a cost of

$$\bar{J}_T = \frac{1}{T} \sum_{t=1}^{T} \left\{ WL_t + \alpha K_t L_{t+1} \mathbb{E} \left[ (s_t - \hat{s}_t)^2 \right] \right\},$$

where we set $R_T = 0, f_T = 0$ for the definition of $\hat{s}_T$.

**A. Lower Bound**

By substituting the result of Th. 2 into Lem. 1 we attain the following lower bound for the achievable LQR cost, which extends the result of (28) to the case of random-rate budgets.

**Theorem 7** (LQR cost lower bound). The optimal LQR cost (30) with rate tuple $R^T$ is bounded from below by

$$\bar{J}_T \geq \frac{1}{T} \sum_{t=1}^{T} \left\{ WL_t + \alpha K_t L_{t+1} D_t^* \right\},$$  \hspace{1cm} \text{(31)}$$

where $K_t$ and $L_t$ are given in Lem. 1 and $D_t^*$ is given in (18).

**Proof:** The proof is immediate by noting that, similarly to the Performance Analysis of Sec. III-A, at time $t$, given $f_t$, all the past control actions $u^{t-1}$ — being a deterministic function of $f^{t-1}$ — can be absorbed into $\hat{s}_t$.

**B. Variable-Length Coding**

Similarly to the proof of Th. 8, by combining the results of Th. 6 and Lem. 1 we attain the following upper bound for the achievable LQR cost, in the VLC scenario; following the exposition in Sec. VII we concentrate here on the packet-eras channel.

**Theorem 8** (VLC LQR cost upper bound). The LQR cost (30) for the VLC scenarios under an average-rate constraint $R$ (25), is bounded from above by

$$\bar{J}_T \leq \frac{1}{T} \sum_{t=1}^{T} \left\{ WL_t + \alpha K_t L_{t+1} D_t \right\},$$  \hspace{1cm} \text{(32)}$$

where $K_t$ and $L_t$ are given in Lem. 7 and $D_t$ is bounded from above as in (29).

**C. Steady State**

We consider here the fixed-parameter fixed-rate case:

$$Q_t \equiv Q, \hspace{1cm} \text{(33a)}$$

$$R_t \equiv R, \hspace{1cm} \text{(33b)}$$

$$R_t \equiv R, \hspace{1cm} \text{(33c)}$$

In case $R_T = 0$, define $K_T = 0$.\[8\]
and similarly to the steady-state distortion \( D \) and average-stage steady-state distortion \([17]\), we wish to determine the optimal steady-state average-stage cost

\[
\bar{J}_\infty \triangleq \limsup_{T \to \infty} \bar{J}_T .
\]

**Corollary 7** (LQR cost lower bound). The steady-state LQR cost for the fixed-parameter fixed-rate case \([33]\) is bounded from below by

\[
\bar{J}_T \geq W L_\infty + \alpha K_\infty L_\infty D_\infty^*, \tag{34}
\]

where \( D_\infty^* \) is given in \([19]\).

\[
K_\infty = \frac{L_\infty}{R + L_\infty} \alpha, \tag{35}
\]

and \( L_\infty \) is the positive solution of

\[
L_\infty^2 - \left[ (\alpha^2 - 1) R + Q \right] L_\infty - QR = 0. \tag{36}
\]

**Remark 14.** As noted in Sec. VII the result of Cor. 8 holds true for both fixed- and variable-length coding.

**Remark 15** (Comparison to separation-based bounds). In \([28]\), it is shown that the optimal steady-state LQR cost must satisfy \([37]\) with the distortion \( D_\infty^* \) dictated by the source–channel separation between the causal rate–distortion \( R_C(D_\infty^*) \) \([6], [9]\) and the directed capacity (maximal directed information) \([43]\). Since in our case the directed capacity is upper bounded by the regular capacity of the channel, \( C = \mathbb{E}[r_1] \), and the causal rate–distortion function (which is in itself a lower bound) is given by \([6], [10]\),

\[
R_C(D_\infty^*) = \frac{1}{2} \log \left( \alpha^2 + \frac{W}{D_\infty^*} \right),
\]

the source–channel separation-based bound \( R_C(D_\infty^*) \leq C \) reduces to the expression in \([19]\) with \( B \triangleq \mathbb{E}[2^{-2r_1}] \) replaced with \( B_{\text{sep}} \triangleq 2^{-2\mathbb{E}[r_1]} \). By applying Jensen’s inequality we see that \( B < B_{\text{sep}} \) for any non-deterministic rate budget distribution. Thus, the joint source and channel treatment offered in this work strengthens the separation-based adaptation of the results in \([28]\). The difference becomes especially pronounced in the packet-erasure and instantaneous ACKs scenario of Sec. [V-A] with an infinite transmission rate \( R \) [recall \([20]\)] — a setting extensively studied in past decade \([17], [44], [45]\). In this case, \( B_{\text{sep}}, \) and consequently also the lower bound on \( D_\infty^* \), reduces to the trivial zero bound, whereas \( B = 1 - \beta > 0 \) unless \( \beta = 1 \).

**Corollary 8** (VLC LQR cost upper bound). The steady-state LQR cost for the packet-erasure fixed-parameter case under an average rate constraint \( R \) \([25]\) is bounded from above by

\[
\bar{J}_T \leq W L_\infty + \alpha K_\infty L_\infty D_\infty^*, \tag{37}
\]

where \( D_\infty, K_\infty, L_\infty \) are given in \([29], [30], [35]\), respectively.

IX. DISCUSSION

A. ACKs with Larger Delays

To extend the delayed ACK scheme of Sec. VI to larger delays, a generalization of Th. 3 is needed. Unfortunately, the optimal rate–distortion region for more than two SI options (e.g., with or without correlated SI \( y \)) remains an open problem and is only known for the (“degraded”) case when the state and the possible SIs form a Markov chain. Nonetheless, achievable regions for multiple SI options have been proposed in \([35]\), which can be used for the construction of schemes that accommodate larger delays.

B. Scalar Fixed-length Coding

In this paper we derived lower bounds and proved that they are tight in the limit of large values of \( N \). In the case of scalar fixed-length quantization, both design and analysis of good schemes are more involved and remain beyond the scope of this paper. For a treatment of the case of log-concave noise distributions (Gaussian included), see \([46]\).

C. Prefix-Free Versus One-Shot Lossless Compression

The VLC ECDQ-based schemes throughout this work employed one-to-one lossless coding. This is a reasonable assumption since, in packeted communications, the descriptions of subsequent symbols may be assumed to be parsed by the underlying protocol, which allows, in turn, to part with the prefix-free constraint and attain better performance \([47]\). Specifically, the bit loss with respect to the entropy of the process of prefix-free coding is circumvented by one-to-one coding \([48]\). Nonetheless, the results of this paper can be easily adjusted to the prefix-free coding case by adding an extra bit on the right hand side of \([25]\) — the maximal loss of prefix-free entropy coding above the entropy, and replacing the factor \( 2\pi e/12 \) in \([27]-[29]\) by \( 2\pi e/3 \).

D. Packet-Erasure Modeling

In this work we modeled the packet-erasure constraint by an i.i.d. process. Nonetheless, the derived results can be extended far beyond this setting, as is evident from the proof of Th. 2.

In the VLC setting, the erasure probability is likely to be higher for longer packets, and calls for further investigation.

E. Non-Gaussian

Following Assert. 1 the lower bounds in this work can be extended to the case of a non-Gaussian driving process \( w_t \), in a straightforward fashion, with the variance of the elements of \( w_t \) in \([15]\) replaced by its entropy power (recall that the two are equal in the Gaussian case), resulting in lower bounds reminiscent of those in \([28]\).

F. Dynamic Rate Allocation over Packet-Erasure Channels

In the packet-erasure setup of Sec. [V-A] a fixed transmission rate \( R \) (uniform rate allocation) was assumed for all time steps. Nonetheless, since at time \( t \) the entire history of the packet erasures \( b_t^{-1} \) is readily available, dynamic rate allocation can be employed, that depends on \( b_t^{-1} \) (which also determines \( D_t^{-1} \)). This setup can be thought of as the source-coding dual of the fast fading channel coding problem where the fading coefficient is known at both the transmitter and the receiver prior to transmission, and the transmitter optimizes the transmission rate via waterfilling across time \([33]\) Ch. 5.4].