# From Ordinary AWGN Codes to Optimal MIMO Wiretap Schemes 

Anatoly Khina, Tel Aviv University

Joint work with:
Yuval Kochman, Hebrew University Ashish Khisti, University of Toronto

ITW 2014<br>Hobart, Tasmania, Australia<br>November 05, 2014

## Channel Model: Gaussian MIMO Wiretap Channel


$\mathbf{y}_{B}=\mathbf{H}_{B} \mathbf{x}+\mathbf{z}_{B}$

$$
\mathbf{y}_{E}=\mathbf{H}_{E} \mathbf{x}+\mathbf{z}_{E}
$$

- $\mathrm{x}-N_{A} \times 1$ input vector of power $P$
- $\mathbf{y}_{B}, \mathbf{y}_{E}-N_{B} \times 1, N_{E} \times 1$ received vectors
- $\mathrm{H}_{B}, \mathrm{H}_{E}-N_{B} \times N_{A}, N_{E} \times N_{A}$ channel matrices
- $\mathbf{z}_{B} \sim \mathcal{C N}\left(\mathbf{0}, \mathbf{I}_{N_{B}}\right), \mathbf{z}_{E} \sim \mathcal{C N}\left(\mathbf{0}, \mathbf{I}_{N_{E}}\right)$ - noise vectors
- "Closed loop" (full channel knowledge everywhere)


## Capacity

## Gaussian SISO channel capacity [Leung-Yan-Cheong, Hellman '78]

$$
C_{S}\left(h_{B}, h_{E}\right)=[\overbrace{\log \left(1+\left|h_{B}\right|^{2} P\right)}^{I\left(X_{i} Y_{B}\right)}-\overbrace{\log \left(1+\left|h_{E}\right|^{2} P\right)}^{I\left(X_{i} Y_{E}\right)}]_{+}
$$

Gaussian MIMO channel capacity [Khisti,Wornell '10][Oggier,Hassibi '11]
$C_{S}\left(\mathbf{H}_{B}, \mathbf{H}_{E}\right)=\max _{\mathbf{K}: \operatorname{trace}\{\mathbf{K}\} \leq P}[\overbrace{\log \left|\mathbf{I}+\mathbf{H}_{B} \mathbf{K} \mathbf{H}_{B}^{\dagger}\right|}^{I\left(\mathbf{X}_{;} \mathbf{Y}_{B}\right)}-\overbrace{\log \left|\mathbf{I}+\mathbf{H}_{E} \mathbf{K H}_{E}^{\dagger}\right|}^{I\left(\mathbf{X}_{\mathbf{\prime}} \mathbf{Y}_{E}\right)}]$

- Maximization over all admissible covariance matrices K
- Power constraint can be replaced with covariance constraint [Liu, Shamai '09]


## How to Construct a Practical Capacity-achieving Scheme?

## Black box approach

- Construct MIMO Wiretap Codes from "ordinary" SISO ones
- Any good "ordinary" SISO AWGN codes
- Signal processing (SVD-based scheme [Telatar '99], V-BLAST [Foschini '96], ...)
- Codeword indexing
- Achieves capacity
- Gap-to-capacity dictated by gap-to-capacity of the SISO codes


## How to Construct a Practical Capacity-achieving Scheme?

## Two-step procedure

(1) Reduce MIMO to SISO (as in "ordinary" MIMO case)
(2) Transform "ordinary" (non-secrecy) codes to wiretap ones

## Weak/strong secrecy

- Concentrate on achievability of weak secrecy
- One specific structure achieves strong secrecy


## "Ordinary" Codes $\rightarrow$ Wiretap Codes

## Good Wiretap Codes for SISO

## Two-level AWGN code of rates $(R, \tilde{R})$

- $x^{n}=g(m, f)$
- $m \in\left\{1, \ldots, 2^{n R}\right\}$ - Information message
- $f \in\left\{1, \ldots, 2^{n \tilde{R}}\right\}$ - Fictitious message
- $g$ - Mapping known to all (including Eve!)
- Bob can decode $(m, f)$ and then discard $f$
- Eve can recover $f$ from $\left(y_{E}, m\right)$ $\Downarrow$
Eve cannot recover $m$ from $y_{E}: I\left(m ; y_{E}\right) \leq n \epsilon$


## Ordinary Codes $\rightarrow$ Two-level AWGN Codes

## Randomized procedure

- Base AWGN codebook $\mathcal{C}_{0}$ of rate $R_{0}: R+\tilde{R}<R_{0}<C_{B}$
- $\forall(m, f)$ : Draw an index $\theta(m, f) \in \operatorname{Unif}\left(\left\{1, \ldots, 2^{n R_{0}}\right\}\right)$
- Average codebook = good two-level AWGN codebook
- De-mapping of random indexing is hard!


## Practical procedure

- Two-universal hash function
[Hayashi, Matsumoto 2010][Bellare, Tessaro, Vardy 2012]
- Low-complexity structured approach
- Valid for Gaussian channels [Tyagi, Vardy ISIT2014]


## MIMO Without Secrecy (No Eve)

## Singular-Value Decomposition (SVD) Scheme [Telatar '99]

- $\mathbf{H}_{B}=\mathbf{Q}_{B} \mathbf{D}_{B} \mathbf{V}_{A}^{\dagger}$
- $\mathbf{Q}_{B}$ and $\mathbf{V}_{A}$ - unitary
- Alice applies $\mathbf{V}_{A}$ and Bob applies $\mathbf{Q}_{B}$
- $\mathbf{D}_{B}=\left(\begin{array}{ccccc}d_{1} & 0 & 0 & \cdots & 0 \\ 0 & d_{2} & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & \cdots & 0 & d_{N-1} & 0 \\ 0 & \cdots & 0 & 0 & d_{N}\end{array}\right) \Rightarrow \begin{gathered}y_{1}=d_{1} x_{1}+z_{1} \\ y_{2}=d_{2} x_{2}+z_{2} \\ \vdots \\ y_{N}=d_{N} x_{N}+z_{N}\end{gathered}$
- Results in parallel scalar sub-channels (each sub-channel has a different SNR)
- Apply water-filling to $\left\{x_{1}, \ldots, x_{N}\right\}: \mathbf{x}=\mathbf{V}_{A} \mathbf{W} \mathbf{c}$


## SVD-based scheme for a given input covariance K

- $\mathbf{H}_{B} \mathbf{K}^{1 / 2}=\mathbf{Q}_{B} \mathbf{D}_{B} \mathbf{V}_{A}^{\dagger}$
- $\mathbf{Q}_{B}$ and $\mathbf{V}_{A}$ - unitary
- Alice applies $\mathrm{K}^{1 / 2} \mathbf{V}_{A}$ and Bob applies $\mathbf{Q}_{B}$
- $\mathbf{D}_{B}=\left(\begin{array}{ccccc}d_{1} & 0 & 0 & \cdots & 0 \\ 0 & d_{2} & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & \cdots & 0 & d_{N-1} & 0 \\ 0 & \cdots & 0 & 0 & d_{N}\end{array}\right) \Rightarrow \begin{gathered}y_{1}=d_{1} x_{1}+z_{1} \\ y_{2}=d_{2} x_{2}+z_{2} \\ \vdots \\ y_{N}=d_{N} x_{N}+z_{N}\end{gathered}$
- Results in parallel scalar sub-channels (each sub-channel has a different SNR)
- Apply filling to $\left\{x_{1}, \ldots, x_{N}\right\}: x=V_{A} W \mathbf{N} \boldsymbol{x}=K^{1 / 2} \mathbf{V}_{A} \mathbf{C}$


## SVD-based scheme for a given input covariance K

- SVD scheme with given K achieves : $R=\log \left|\mathbf{I}_{N_{A}}+\mathbf{H}_{B} \mathrm{KH}_{B}^{\dagger}\right|$
- For optimal choice of $K$ attains capacity
- Can be used to attain capacity for other covariance constraint scenarios (e.g., individual power constraints)


## V-BLAST Scheme: QR Decomposition Based Scheme

## Zero-forcing V-BLAST [Foschini '96][Wolniansky et al. '98]

- $\mathbf{H}_{B}=\mathbf{Q}_{B} \mathbf{T}_{B}$
- $\mathbf{Q}_{B}$ - unitary; $\mathbf{T}_{B}$ - triangular
- Bob applies $\mathbf{Q}_{B}^{\dagger}$ (no SP is required by Alice)
$-\mathbf{T}_{B}=\left(\begin{array}{ccccc}t_{1} & * & * & \cdots & * \\ 0 & t_{2} & * & \cdots & * \\ \vdots & \vdots & \ddots & \ddots & \vdots \\ 0 & 0 & \cdots & t_{N-1} & * \\ 0 & 0 & \cdots & 0 & t_{N}\end{array}\right) \quad \Rightarrow \quad \begin{gathered}y_{1}^{\mathrm{eff}}=t_{1} x_{1}+z_{1} \\ y_{2}^{\mathrm{eff}}=t_{2} x_{2}+z_{2} \\ \vdots \\ \\ y_{N}^{\mathrm{eff}}=t_{N} x_{N}+z_{N}\end{gathered}$
- Off-diagonal elements are canceled via successive interference cancellation (SIC)


## V-BLAST Scheme: QR Decomposition Based Scheme

MMSE-VBLAST for a given covariance K [Hassibi '00]

- $\left[\begin{array}{c}\mathbf{H}_{B} \mathbf{K}^{1 / 2} \\ \mathbf{I}_{N_{A}}\end{array}\right]=\mathbf{Q}_{B} \mathbf{T}_{B}$
- $\mathbf{Q}_{B}$ - unitary; $\tilde{\mathbf{Q}}_{B}-N_{B} \times N_{A}$ submatrix of $\mathbf{Q}_{B}$
- Bob applies $\tilde{\mathrm{Q}}_{B}^{\dagger}$ (no SP is required by Alice)
- $\tilde{\mathbf{Q}}_{B}^{\dagger}$ contains Wiener-filtering ("FFE")
- Effective noise has channel noise and "ISI" components
- Effective SNRs satisfy: $t_{i}^{2}=1+$ SNR $_{i}$

$$
\log \left(t_{i}^{2}\right)=\log \left(1+\operatorname{SNR}_{i}\right)=I\left(c_{i} ; \mathbf{y}_{B} \mid c_{i+1}^{N_{A}}\right)
$$

- Off-diagonal elements above diagonal canceled via SIC


## V-BLAST Scheme: QR Decomposition Based Scheme

- For square invertible $\mathbf{H}$, ZF-VBLAST achieves:

$$
R=\log \left|\mathbf{H}_{B} \mathbf{H}_{B}^{\dagger}\right|
$$

(Using K at the transmitter achieves: $R=\log \left|\mathbf{H}_{B} \mathbf{K H}_{B}^{\dagger}\right|$ )

- MMSE-VBLAST achieves: $R=\log \left|\mathbf{I}_{N_{B}}+\mathbf{H}_{B} \mathbf{K} \mathbf{H}_{B}^{\dagger}\right|$


## Precoded V-BLAST

MMSE-VBLAST with precoding for a given covariance K
$\cdot\left[\begin{array}{c}\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\ \mathbf{I}_{N_{A}}\end{array}\right]=\mathbf{Q}_{B} \mathbf{T}_{B}$

- $\mathrm{V}_{\mathrm{A}}$ can be used to design diagonal values $\Leftrightarrow$ design SNRs


## Precoded V-BLAST

MMSE-VBLAST with precoding for a given covariance K
$-\left[\begin{array}{c}\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\ \mathbf{I}_{N_{A}}\end{array}\right]=\mathbf{Q}_{B} \mathbf{T}_{B}$

- $\mathrm{V}_{\mathrm{A}}$ can be used to design diagonal values $\Leftrightarrow$ design SNRs


## SVD-scheme as MMSE-VBLAST (QR)

Choosing $\mathrm{V}_{A}$ of the SVD of $\mathrm{H}_{B} \mathrm{~K}^{1 / 2} \Rightarrow$ SVD scheme (no SIC needed)

## Precoded V-BLAST

MMSE-VBLAST with precoding for a given covariance K
$-\left[\begin{array}{c}\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\ \mathbf{I}_{N_{A}}\end{array}\right]=\mathbf{Q}_{B} \mathbf{T}_{B}$

- $\mathbf{V}_{A}$ can be used to design diagonal values $\Leftrightarrow$ design SNRs

SVD-scheme as MMSE-VBLAST (QR)
Choosing $\mathbf{V}_{A}$ of the SVD of $\mathbf{H}_{B} \mathrm{~K}^{1 / 2} \Rightarrow$ SVD scheme (no SIC needed)

Geometric-mean decomposition [Jiang et al. '05]/ QRS [Zhang et al. '05]

- $\mathbf{V}_{A}$ is choosing s.t. all diagonal values (all SNRs) are equal
- The same codebook can be used over all subchannels
- No need for bit-loading


## V-BLAST: What Codes Can be Used?

## Problem

- Not any codebooks can be used!
- At each stage of V-BLAST: Noise $=$ Gaussian noise + ISI
- Aligned codes impair decoding


## Alignment phenomenon

For the decoding of sub-stream $x_{i}$

- Bob Cancels out $x_{i+1}, \ldots, x_{N}$
- Applies maximum ratio combining for the recovery of $x_{i}$
- Example: Suppose the resulting effective channel is

$$
y_{i}^{\mathrm{eff}}=2 x_{i}+\underbrace{x_{i-1}+z_{i}}_{\text {Effective noise }}
$$

- If $x_{i}, x_{i-1}$ belong to same lattice codebook $\Rightarrow 2 x_{i}+x_{i-1}$ is not uniquely decodable!


## V-BLAST: What Codes Can be Used?

- In V-BLAST: Bob observes a MAC channel at each stage $i$


## Multiple-access (MAC) SIC codes

- A collection of AWGN codes that are "sufficiently different"
- No MAC gains can align them
- Relaxation of the "MAC capacity-achieving codes" of [Baccelli, El Gamal, Tse 2011]

How to generate such codes?
Theoretical: Encapsulate in dithered modulo lattice of high dim.

- Not black box! ${ }^{-}$

Practical: Simple randomization process suffice (not rigor!):

- Multiplicative (phase) dithering
- Different interleaving / permutation of each code


## Putting It All Together

## New Scheme for General SNR [Kh., Kochman, Khisti ISIT2014]

$$
\begin{aligned}
& {\left[\begin{array}{c}
\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{B} \overbrace{\left(\begin{array}{ccc}
b_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & b_{N}
\end{array}\right)}^{\mathbf{T}_{B}}, \quad b_{i}^{2}=1+\mathrm{SNR}_{i}^{B}} \\
& {\left[\begin{array}{c}
\mathbf{H}_{E} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{E} \overbrace{\left(\begin{array}{ccc}
e_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & e_{N}
\end{array}\right)}, \quad e_{i}^{2}=1+\mathrm{SNR}_{i}^{E}}
\end{aligned}
$$

- Use good SISO wiretap codes for SNR-pairs ( $\left.b_{i}^{2}-1, e_{i}^{2}-1\right)$


## New Scheme for General SNR [Kh., Kochman, Khisti ISIT2014]

$$
\begin{aligned}
& {\left[\begin{array}{c}
\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{B} \overbrace{\left(\begin{array}{ccc}
b_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & b_{N}
\end{array}\right)}^{\mathbf{T}_{B}}, \quad b_{i}^{2}=1+\mathrm{SNR}_{i}^{B}} \\
& {\left[\begin{array}{c}
\mathbf{H}_{E} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{E} \overbrace{\left(\begin{array}{ccc}
e_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & e_{N}
\end{array}\right)}^{\mathbf{T}_{E}}, \quad e_{i}^{2}=1+\mathrm{SNR}_{i}^{E}}
\end{aligned}
$$

- Use good SISO wiretap codes for SNR-pairs $\left(b_{i}^{2}-1, e_{i}^{2}-1\right)$
- $\mathbf{V}_{A}$ of Eve's SVD $\Rightarrow$ Easy secrecy analysis + strong secrecy


## New Scheme for General SNR [Kh., Kochman, Khisti ISIT2014]

$$
\begin{aligned}
& {\left[\begin{array}{c}
\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{B} \overbrace{\left(\begin{array}{ccc}
b_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & b_{N}
\end{array}\right)}^{\mathbf{T}_{B}}, \quad b_{i}^{2}=1+\mathrm{SNR}_{i}^{B}} \\
& {\left[\begin{array}{c}
\mathbf{H}_{E} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{E} \overbrace{\left(\begin{array}{ccc}
e_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & e_{N}
\end{array}\right)}, \quad e_{i}^{2}=1+\mathrm{SNR}_{i}^{E}}
\end{aligned}
$$

- Use good SISO wiretap codes for SNR-pairs ( $b_{i}^{2}-1, e_{i}^{2}-1$ )
- $\mathbf{V}_{A}$ of Eve's SVD $\Rightarrow$ Easy secrecy analysis + strong secrecy
- $\mathbf{V}_{A}$ of Bob's SVD $\Rightarrow$ No need for V-BLAST


## New Scheme for General SNR [Kh., Kochman, Khisti ISIT2014]

$$
\begin{aligned}
& {\left[\begin{array}{c}
\mathbf{H}_{B} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{B} \overbrace{\left(\begin{array}{ccc}
b_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & b_{N}
\end{array}\right)}^{\mathbf{T}_{B}}, \quad b_{i}^{2}=1+\mathrm{SNR}_{i}^{B}} \\
& {\left[\begin{array}{c}
\mathbf{H}_{E} \mathbf{K}^{1 / 2} \mathbf{V}_{A} \\
\mathbf{I}_{N_{A}}
\end{array}\right]=\mathbf{Q}_{E} \overbrace{\left(\begin{array}{ccc}
e_{1} & * & * \\
0 & \ddots & * \\
0 & 0 & e_{N}
\end{array}\right)}, \quad e_{i}^{2}=1+\mathrm{SNR}_{i}^{E}}
\end{aligned}
$$

- Use good SISO wiretap codes for SNR-pairs ( $b_{i}^{2}-1, e_{i}^{2}-1$ )
- $\mathbf{V}_{A}$ of Eve's SVD $\Rightarrow$ Easy secrecy analysis + strong secrecy
- $\mathbf{V}_{A}$ of Bob's SVD $\Rightarrow$ No need for V-BLAST
- $\operatorname{diag}\left\{T_{B}\right\}, \operatorname{diag}\left\{T_{E}\right\}$ are const. $\Rightarrow$ Same code over all channels


## New Scheme for General SNR [Kh., Kochman, Khisti ISIT2014]

## But...

- Proof used random binning $\Rightarrow$ Existence result


## New Scheme for General SNR [Kh., Kochman, Khisti ISIT2014]

## But...

- Proof used random binning $\Rightarrow$ Existence result

Theorem
Good two-level MAC-SIC codes approach the MIMO WTC capacity.

## Two-Level MAC-SIC Codes Achieve MIMO WTC Capacity

## Proof idea

- Bob's optimal (?) receiver of sub-message $i$ :
- Sub-messages $(i+1), \ldots, N$ are known
- Subtract interference of $x_{i+1}, \ldots, x_{N}$
- Treat $x_{1}, \ldots, x_{i-1}$ as noise
- Project onto subspace of $x_{i}$
- Eve's genie-aided optimal (?) receiver of sub-message $i$ :
- Sub-messages $(i+1), \ldots, N$ are revealed to Eve for decoding $x_{i}$
- Subtract interference of $x_{i+1}, \ldots, x_{N}$
- Treat $x_{1}, \ldots, x_{i-1}$ as noise
- Project onto subspace of $x_{i}$
- Secrecy: Codes need to be two-level
- Optimality: Codes need to be MAC-SIC


## End-to-End Scheme

## "Nested black-box" type approach

## End-to-End Scheme

## "Nested black-box" type approach

or

## "Матрёшка" ("Matryoshka") type approach



## End-to-End Scheme

## Modulation

- Apply the MIMO wiretap matrix decomposition scheme
- Bob uses standard V-BLAST for decoding


## End-to-End Scheme

## Modulation

- Apply the MIMO wiretap matrix decomposition scheme


## Coding: Good two-level MAC-SIC codes

- Take any good AWGN codes of appropriate rates $\left\{R_{i}+\tilde{R}_{i}\right\}$
- Transform into "good MAC-SIC codes" via a randomization process (modulo-lattice, interleaving,...)
- Transform into "good two-level codes" via random indexing / two-universal hashing
- Bob uses standard V-BLAST for decoding


## End-to-End Scheme

## Modulation

- Apply the MIMO wiretap matrix decomposition scheme


## Coding: Good two-level MAC-SIC codes

- Take any good AWGN codes of appropriate rates $\left\{R_{i}+\tilde{R}_{i}\right\}$
- Transform into "good MAC-SIC codes" via a randomization process (modulo-lattice, interleaving,...)
- Transform into "good two-level codes" via random indexing / two-universal hashing
- Bob uses standard V-BLAST for decoding

Alignment has a double-bad effect in wiretap

- Bob cannot recover the whole message
- ISI that serves as noise for Eve might align


## Complementary

## Good Wiretap Codes for SISO

## Explanation of last requirement

$$
\left.\begin{array}{rl}
I\left(x^{n} ; y_{E}^{n}\right) & =I\left(m, f ; y_{E}^{n}\right) \\
=I\left(m ; y_{E}^{n}\right)+I\left(f ; y_{E}^{n} \mid m\right) \\
& =\underbrace{H\left(m ; y_{E}^{n}\right)+}-\underbrace{H(f)}_{\leq n \delta_{2}} \leq n\left(y_{E}^{n}, m\right)
\end{array} n C_{E}\right)
$$

