# Error probability bounds in information theory:

Role of structure, performance criteria and decision rules

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January 21, 2018

### Outline

- Introduction: Error exponent for single user channel
- Overview of linear codes in network problems
- Contribution I: Distributed expurgation using structured codes for network problems – terminals use different linear codes
- Contribution II: Distributed hypothesis testing using structured codes – terminals use **same** linear code

# Single-User Channel

Transmitter 
$$x_1, \dots, x_n$$
  $p(y|x)$   $y_1, \dots, y_n$  Receiver

Memoryless channel

$$p(y_1,\ldots,y_n|x_1,\ldots,x_n) = \prod_{t=1}^n p(y_t|x_t)$$

#### **Basic definitions:**

- Blocklength n: number of channel uses
- Codebook C: a set of  $M = 2^{nR}$  codewords (vectors of length *n*)
- Average Error Probability: P<sub>e</sub> = ℙ(Ĉ ≠ C), where C ~ Uniform (C)

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#### Basic tradeoff:

Tradeoff between number of codewords, blocklength and average error probability

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#### First-Order (Capacity): asymptotics in blocklength

Capacity C: Highest achievable rate with vanishing  $P_e$  as  $n \to \infty$ 

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## Shannon Theory

- Random Code: Symbol-wise (and codeword-wise) i.i.d. p(x)
- Information Density:

$$i(X; Y) \stackrel{ riangle}{=} \log rac{p(X, Y)}{p(X)p(Y)}$$

Mutual Information:

$$I(X; Y) \stackrel{\triangle}{=} \mathbb{E}i(X; Y)$$

Shannon's Channel Coding Theorem ['48] (first-order characterization)

 $C = \max_{p(x)} I(X; Y)$ 

maximization over all input distributions p(x)

January 21, 2018

# Tradeoff: Refined Analysis

- There is a long history of finite blocklength bounds: Elias, Feinstein, Gallager, ...
- Polyanskiy et al. [2010] gave two simple achievability bounds (DT & RCU). Disturbing point: neither dominates
- We have resolved this issue (but not in this talk...)
- Asymptotic analysis: the error event amounts to (except for low rates)

$$i(X^n; Y^n) \triangleq \frac{1}{n} \sum_{k=1}^n i(X_k; Y_k) < R$$

# Asymptotic Bounds on the Information Density

The following asymptotics are with respect to the blocklength (for high rates):

- Central Limit Theorem (CLT): good for high P<sub>e</sub>, dispersion [Strassen 1962, Polyanskiy et al. 2010]
- We have derived results regarding the extension to network problems (but not in this talk...)
- Large Deviations Principle (LDP): good for low Pe, exponent

$$\Pr\left\{i(X^n; Y^n) < R\right\} \le \exp\{-nE(R)\}$$

Similar lower bounds are known

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## Error Exponent: Code Structure May Matter

- High rates: typical error due to a "bad" channel *i*(*X<sup>n</sup>*; *Y<sup>n</sup>*) < *R*. Random coding achieves the exponent
- Low rates: typical error due to "bad" codewords (e.g. for BSC, minimum distance dominates)
- Can be solved by expurgation of random codes, or (almost all) linear codes
- Who cares about expurgation? For almost noiseless (binary input) channels

$$\frac{R_{\text{ex}}}{C} \xrightarrow[C \to 1]{} 1$$



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# But First: Why Linear Codes in Single-User Channel?

- Whenever uniform distribution is optimal, linear codes achieve capacity, exponents, dispersion
- But no theoretical gain
- Historically, interest was due to practical (complexity) advantages

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# Why Linear Codes in Networks?

#### Contribution II (in this talk ... )

- Recent interest, reviving a theme introduced by Körner-Marton 1979: first-order (capacity) advantage in some network settings (Nazer & Gastpar, Wilson et al., Philisof et al., ...)
- In this work: distributed hypothesis testing
- Terminals use the same linear code

#### Contribution I (in this talk...)

- Error-probability advantage in network settings (even when no first-order gain) – multiple-access (MAC) channel
- Terminals use different linear codes
- The prospect for such an improvement was hinted to in a distributed source coding context by Csiszár [1982, "Linear Codes for Sources and Source Networks: Error Exponents, Universal Coding"]

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## MAC Channel

• For simplicity 2 users



• Capacity region: the closure of the convex-hull of all  $(R_1, R_2)$  satisfying:  $R_2$ 



over some product distribution  $p(x_1, x_2) = p(x_1)p(x_2)$ 

# Toy Example: Erasure-Additive MAC Channel



#### Obvious bounds on $P_e$

- Lower bound: single-user erasure channel
- Upper bound: same with half blocklength (time sharing)

#### Is any of these bounds tight?

# What Can Be Achieved Using Random Codes?

- Slepian & Wolf ['73], Gallager ['85]
- Receiver's perspective: sum of codebooks,  $C = C_1 + C_2$
- For random codes: summation preserves pairwise independence, thus most standard bounds (RCU, DT, dispersion, random exponent) hold
- Codebook structure (e.g. minimum distance) is not preserved
- But recall that minimum distance dictates error exponent at low rates
- Expurgation attempts recently by Nazari et al.: expurgate one user (even for MAC channel with many users)

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# Solution: Use Linear Codes

- Create a linear sum-codebook (recall: inherently expurgated)
- Simply split the generating matrix between users
- At the receiver, the summation is *indistinguishable* from a single user channel with the sum-rate
- Performance identical to single user with the sum rate
- Any performance that is attainable via linear codes over the single-user channel is also attainable for the considered MAC
- The generation process is equivalent to generating two different linear codes

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# The Error Exponent of MAC Channels

- In toy example: single-user random+expurgated exponents are achievable
- Extends to any MAC channel that is finite-field summation + single-user channel (e.g., BSC MAC)
- Advantage for any "similar" channel (by continuity)
- AWGN MAC channel constraints are a challenge.
- For certain parameters improving on Gallager ['85]
- General case: wide open.



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## Distributed Hypothesis Testing [Berger '79]



 $\mathcal{H}_0 : (\mathbf{X}, \mathbf{Y}) \sim \text{i.i.d. } P_0(x, y)$  $\mathcal{H}_1 : (\mathbf{X}, \mathbf{Y}) \sim \text{i.i.d. } P_1(x, y)$ 

- Rates:  $R_X = 1/n \cdot \log |\mathcal{M}_X|, R_Y = 1/n \cdot \log |\mathcal{M}_Y|$
- Error probabilities  $\{\epsilon_0\}, \{\epsilon_1\}$  as in standard hypothesis testing
- But now, there is a tradeoff between rates, error probabilities and blocklength
- Long history: Ahlswede & Csiszár '81, '86, Han '87, Shalaby & Papamarcou '92, Shimokawa et al. '94, Han & Amari '98, Rahman & Wagner 2012...

# Rate-Exponents Tradeoff

For (a sequence of) error probabilities {*ϵ*<sub>0</sub>(*n*)}, {*ϵ*<sub>1</sub>(*n*)}, the exponential decay rates are defined as:

$$E_i = \liminf_{n \to \infty} -\frac{1}{n} \log \epsilon_i(n)$$

- Goal: Characterize the achievable region of (*E*<sub>0</sub>, *E*<sub>1</sub>) pairs subject to the rate constraints
- Two extreme (and natural) cases:
  - Side-information case: *R*<sub>Y</sub> unconstrained
  - Symmetric rate constraints:  $R_X = R_Y = R$

# **Binary Symmetric Case**

- Under both hypotheses, (X, Y) is a doubly-symmetric source
- Noise / difference sequence: Z = (X + Y) mod 2
- $\mathcal{H}_i$ : *Z* is Bernoulli- $p_i$ , where  $p_0 < p_1 \le 1/2$
- The key point is that the type of Z is a sufficient statistic
- For R ≥ 1, the unconstrained exponents are achievable: For any *p*<sub>0</sub> ≤ *s* ≤ *p*<sub>1</sub>,

$$E_0(s) = D_b(s \| p_0)$$
  
 $E_1(s) = D_b(s \| p_1)$ 

where  $D_b(\cdot)$  is the binary KL divergence

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#### Side-Information Case: Random Binning [Shimokawa et al. '94]

- Base on Slepian-Wolf coding (random binning)
- Decoder recovers the sources first (decoding similar to BSC decoder with "channel" noise Z)
- Key observation: under a binning error, typically the reconstruction will *not* fall in the vicinity of Y
- This gives a non-trivial exponent pair
- Can be improved by using quantization
- We have further improvements using geometric analysis (but not in this talk...)
- But what about the symmetric constraints case?

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# Körner-Marton Reminder



- Setting: Suppose we wish to compress the difference Z = X + Y (X and Y BSS pair) in a distributed manner
  - Using SW (first reconstructing X, Y) requires:

$$R_X = H(\mathbf{Z}), R_Y = H(\mathbf{Y})$$

• But KM showed that it suffices to require:  $R_X = H(\mathbf{Z}), R_Y = H(\mathbf{Z})$ 

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• Again: linear codes are the way to go

# Körner-Marton Coding Scheme (crash course)

- Let H be a parity-check matrix of a linear code of rate R
- $\phi_X(\mathbf{X}) = H\mathbf{X}, \, \phi_Y(\mathbf{Y}) = H\mathbf{Y}$  have rate 1 R
- The decoder evaluates HX + HY = HZ
- Finally, a syndrome decoder is used
- $\hat{\mathbf{Z}} = \mathbf{Z}$  if and only if  $\mathbf{Z}$  is inside the basic "Voronoi" cell
- Same error event as in the side-information (SW coding) case

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# Main Result

#### Achievable tradeoff for symmetric constraints

- We can leverage KM coding to the distributed hypothesis problem
- (Essentially the) same exponents are therefore achievable, as in the side-information case

$$\begin{array}{ccc} {\sf SW} & {\sf Random-binning DHT} \\ \downarrow & \downarrow \\ {\sf KM} & {\sf KM-style DHT} \end{array}$$

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# Thank you for your attention!

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